

[LMV225](http://www.ti.com/product/lmv225?qgpn=lmv225), [LMV226](http://www.ti.com/product/lmv226?qgpn=lmv226), [LMV228](http://www.ti.com/product/lmv228?qgpn=lmv228)

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LMV225/LMV226/LMV228 RF Power Detector for CDMA and WCDMA

Check for Samples: [LMV225,](http://www.ti.com/product/lmv225#samples) [LMV226,](http://www.ti.com/product/lmv226#samples) [LMV228](http://www.ti.com/product/lmv228#samples)

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- -
	- **– DSBGA Ultra Thin 1.0 mm ^x 1.0 mm ^x 0.35** external capacitor. **mm**
	-

-
-
- **CDMA2000 RF Power Control**
-

¹FEATURES DESCRIPTION

The LMV225/LMV226/LMV228 are 30 dB RF power **²• 30 dB Linear in dB Power Detection Range** detectors intended for use in CDMA and WCDMA **• Output Voltage Range 0.2 to 2V** applications. The device has an RF frequency range from 450 MHz to 2 GHz. It provides an accurate **• Multi-Band Operation from 450 MHz to 2000** temperature and supply compensated output voltage **MHz MHz** that relates linearly to the RF input power in dBm. The circuit operates with a single supply from 2.7V to **FIRE COMPETATION FIRE COMPONSATION**
5.5V. The LMV225/LMV226/LMV228 have an
integrated filter for low-ripple average power detection **• Packages:** integrated filter for low-ripple average power detection **– DSBGA Thin 1.0 mm x 1.0 mm x 0.6 mm** of CDMA signals with 30 dB dynamic range. Additional filtering can be applied using a single external capacitor.

– WSON 2.2 mm x 2.5 mm x 0.8 mm The LMV225 has an RF power detection range from –30 dBm to 0 dBm and is ideally suited for direct use
in combination with resistive taps. The **– (LMV225 and LMV228)** in combination with resistive taps. The LMV226/LMV228 have a detection range from –15 **APPLICATIONS** dBm to 15 dBm and are intended for use in combination with a directional coupler. The LMV226 **• CDMA RF Power Control** is equipped with a buffered output which makes it **• WCDMA RF Power Control** suitable for GSM, EDGE, GPRS and TDMA

•• PA Modules ••• PA Modules Fig. 2 PA Modules PA Modules Fig. 2 PA Module 2 PA M a low power consumption shutdown mode. During shutdown the output will be LOW. The output voltage ranges from 0.2V to 2V and can be scaled down to meet ADC input range requirements.

> The LMV225/LMV226/LMV228 power detectors are offered in the thin 1.0 mm x 1.0 mm x 0.6 mm DSBGA package and the ultra thin 1.0 mm x 1.0 mm x 0.35 mm DSBGA package. The LMV225 and the LMV228 are also offered in the 2.2 mm x 2.5 mm x 0.8 mm WSON package.

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These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

ABSOLUTE MAXIMUM RATINGS (1)(2)

(1) Absolute Maximum Ratings indicate limits beyond which damage to the device may occur. Operating Ratings indicate conditions for which the device is intended to be functional, but specific performance is not specified. For specifications and the test conditions, see the Electrical Characteristics.

(2) If Military/Aerospace specified devices are required, please contact the Texas Instruments Sales Office/ Distributors for availability and specifications.

- (3) Human body model: 1.5 kΩ in series with 100 pF. Machine model, 0Ω in series with 100 pF.
- (4) The maximum power dissipation is a function of $T_{J(MAX)}$, θ_{JA} and T_A . The maximum allowable power dissipation at any ambient temperature is P_D = (T_{J(MAX)} - T_A)/θ_{JA}. All numbers apply for packages soldered directly into a PC board

OPERATING RATINGS (1)

(1) Absolute Maximum Ratings indicate limits beyond which damage to the device may occur. Operating Ratings indicate conditions for which the device is intended to be functional, but specific performance is not specified. For specifications and the test conditions, see the Electrical Characteristics.

2.7 DC AND AC ELECTRICAL CHARACTERISTICS

Unless otherwise specified, all limits are specified to $V_{DD} = 2.7V$; $T_J = 25^{\circ}C$. **Boldface** limits apply at temperature extremes. ⁽¹⁾

(1) Electrical Table values apply only for factory testing conditions at the temperature indicated. Factory testing conditions result in very limited self-heating of the device such that $T_J = T_A$. No specification of parametric performance is indicated in the electrical tables under

conditions of internal self-heating where $T_J > T_A$.
(2) All limits are specified by design or statistical analysis

(3) Turn-on time is measured by connecting a 10 kΩ resistor to the RF_{IN}/E_N pin. Be aware that in the actual application on the front page, the RC-time constant of resistor R_2 and capacitor C adds an additional delay.

(4) Typical values represent the most likely parametric norm.

(5) Power in dBV = dBm + 13 when the impedance is 50 Ω .

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STRUMENTS

EXAS

2.7 DC AND AC ELECTRICAL CHARACTERISTICS (continued)

Unless otherwise specified, all limits are specified to V_{DD} = 2.7V; T_J = 25°C. **Boldface** limits apply at temperature extremes. [\(1\)](#page-7-0)

(6) Device is set in active mode with a 10 kΩ resistor from V_{DD} to RF_{IN}/E_N. RF signal is applied using a 50Ω RF signal generator AC coupled to the RF_{IN}/E_N pin using a 100 pF coupling capacitor.

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STRUMENTS

2.7 DC AND AC ELECTRICAL CHARACTERISTICS (continued)

Unless otherwise specified, all limits are specified to V_{DD} = 2.7V; T_J = 25°C. Boldface limits apply at temperature extremes. [\(1\)](#page-7-0)

STRUMENTS

EXAS

5.0 DC AND AC ELECTRICAL CHARACTERISTICS

Unless otherwise specified, all limits are specified to $V_{DD} = 5.0V$; $T_J = 25^{\circ}C$. **Boldface** limits apply at temperature extremes. ⁽¹⁾

(1) Electrical Table values apply only for factory testing conditions at the temperature indicated. Factory testing conditions result in very limited self-heating of the device such that $T_J = T_A$. No specification of parametric performance is indicated in the electrical tables under conditions of internal self-heating where $T_J > T_A$.

(2) All limits are specified by design or statistical analysis

(3) Turn-on time is measured by connecting a 10 kΩ resistor to the RF_{IN}/E_N pin. Be aware that in the actual application on the front page, the RC-time constant of resistor R_2 and capacitor C adds an additional delay.

(4) Typical values represent the most likely parametric norm.

(5) Power in dBV = dBm + 13 when the impedance is 50 Ω .

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5.0 DC AND AC ELECTRICAL CHARACTERISTICS (continued)

Unless otherwise specified, all limits are specified to $V_{DD} = 5.0V$; $T_J = 25^{\circ}C$. **Boldface** limits apply at temperature extremes. [\(1\)](#page-7-0)

(6) Device is set in active mode with a 10 kΩ resistor from V_{DD} to RF_{IN}/E_N. RF signal is applied using a 50Ω RF signal generator AC coupled to the RF_{IN}/E_N pin using a 100 pF coupling capacitor.

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5.0 DC AND AC ELECTRICAL CHARACTERISTICS (continued)

Unless otherwise specified, all limits are specified to $V_{DD} = 5.0V$; $T_J = 25^{\circ}C$. **Boldface** limits apply at temperature extremes. [\(1\)](#page-7-0)

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CONNECTION DIAGRAM

Figure 3. 4-Bump DSBGA – Top View Figure 4. 6-pin WSON – Top View See Package Number YZR0004 or YPD0004 See Package Number NGF0006A

PIN DESCRIPTIONS

Block Diagrams

Figure 7. LMV228

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TYPICAL PERFORMANCE CHARACTERISTICS LMV225

Unless otherwise specified, $V_{DD} = 2.7V$, $T_J = 25^{\circ}C$.

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ISTRUMENTS

(continued)

Logarithmic Slope vs. Frequency (LMV225) Logarithmic Intercept vs. Frequency (LMV225)

Output Variation vs. RF Input Power Output Variation vs. RF Input Power
Normalized to 25°C @ 900 MHz (LMV225) Normalized to 25°C @ 1800 MHz (LMV225)

Output Variation vs. RF Input Power Output Variation vs. RF Input Power
Normalized to 25°C @ 1900 MHz (LMV225) Normalized to 25°C @ 2000 MHz (LMV225)

TYPICAL PERFORMANCE CHARACTERISTICS LMV225

TYPICAL PERFORMANCE CHARACTERISTICS LMV226

Unless otherwise specified, $V_{DD} = 2.7V$, $T_J = 25^{\circ}C$.

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TYPICAL PERFORMANCE CHARACTERISTICS LMV226 (continued)

Unless otherwise specified, $V_{DD} = 2.7V$, $T_{J}= 25^{\circ}C$.
Output Voltage and Log Conformance vs.
RF Input Power @ 900 MHz (LMV226)

Output Voltage and Log Conformance vs. Output Voltage and Log Conformance vs. RF Input Power
RF Input Power @ 900 MHz (LMV226)

Logarithmic Slope vs. Frequency (LMV226) Logarithmic Intercept vs. Frequency (LMV226)

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ERROR (dB)

ERROR (dB)

-1.5 -1.0 -0.5 0.0

-50 -40 -30 -20 -10 0 10 RF INPUT POWER (dBm)

-40°C

PSRR vs. Frequency RF Input Impedance vs. Frequency @ Resistance and (LMV226) Reactance (LMV226)

TYPICAL PERFORMANCE CHARACTERISTICS LMV228 IN DSBGA

Unless otherwise specified, $V_{DD} = 2.7V$, $T_{J} = 25^{\circ}C$.

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2.50 5 2.25 85°C 4 25°C 2.00 3 85°C 2 1.75 -40° C (dB) ERROR (dB) 25°C 1.50 1 VOUT (V) ERROR 0 1.25 -40°C 1.00 -1 0.75 -2 0.50 -3 0.25 -4 $0.00 - 50$ -5 -50 -40 -30 -20 -10 0 10 20 RF INPUT POWER (dBm)

TYPICAL PERFORMANCE CHARACTERISTICS LMV228 IN DSBGA

(continued)

Unless otherwise specified, $V_{DD} = 2.7V$, $T_{J}= 25^{\circ}C$.

Logarithmic Slope vs.

Frequency (LMV228 in DSBGA)

TYPICAL PERFORMANCE CHARACTERISTICS LMV228 IN DSBGA (continued)

TYPICAL PERFORMANCE CHARACTERISTICS LMV228 IN WSON

Unless otherwise specified, $V_{DD} = 2.7V$, $T_{J} = 25^{\circ}C$.

vs. Output Voltage and Log Conformance vs. RF Input Power @ 900 MHz (LMV228 inWSON) RF Input Power @ 1800 MHz (LMV228 in WSON)

EXAS

STRUMENTS

$$
f_{\rm{max}}
$$

APPLICATION NOTES

CONFIGURING A TYPICAL APPLICATION

The LMV225/LMV226/LMV228 are power detectors intended for CDMA and WCDMA applications. Power applied at its input translates to a DC voltage on the output through a linear-in-dB response. The LMV225 detector is especially suited for power measurements via a high-resistive tap, while the LMV226/LMV228 are designed to be used in combination with a directional coupler. The LMV226 has an additional output voltage buffer and therefore a low output impedance. The key features of the devices are shown in .

In order to match the output power range of the power amplifier (PA) with the range of the LMV225's input, the high resistive tap needs to be configured correctly. In case of the LMV226/LMV228 the coupling factor of the directional coupler needs to be chosen correctly.

HIGH RESISTIVE TAP APPLICATION

The constant input impedance of the device enables the realization of a frequency independent input attenuation to adjust the LMV225's range to the range of the PA. Resistor R₁ and the 50Ω input resistance (R_{IN}) of the device realize this attenuation ([Figure](#page-21-0) 66). To minimize insertion loss, resistor R_1 needs to be sufficiently large. The following example demonstrates how to determine the proper value for R_1 .

Figure 66. Typical LMV225 Application with High Resistive Tap

Suppose the useful output power of the PA ranges up to +31 dBm. As the LMV225 can handle input power levels up to 0 dBm. R₁ should realize a minimum attenuation of 31 - 0 = 31 dB. The attenuation realized by R₁ and the effective input resistance R_{IN} of the detector equals:

$$
A_{dB} = 20 \cdot LOG \left[1 + \frac{R_1}{R_{IN}}\right] = 31dB
$$

Solving this expression for R₁, using that $R_{IN} = 50\Omega$, yields:

$$
R_1 = \left[\begin{array}{c} A_{\text{dB}} \\ 10^{\overline{20}} & 1 \end{array} \right], \ R_{\text{IN}} = \left[\begin{array}{c} 31 \\ 10^{\overline{20}} & 1 \end{array} \right], \ 50 = 1724 \Omega
$$

In [Figure](#page-21-0) 66, R₁ is set to 1800Ω resulting in an attenuation of 31.4 dB.

(2)

(1)

DIRECTIONAL COUPLER APPLICATION

The LMV226/LMV228 also has a 50Ω input resistance. However, its input range differs compared to the LMV225, i.e. −15 dBm to +15 dBm. If a typical attenuation of a directional coupler is 20 dB, the LMV226/LMV228 can be directly connected via the directional coupler to the PA without the need of additional external attenuator [\(Figure](#page-22-0) 67). Different PA ranges can be configured using couplers with other coupling factors.

Figure 67. Typical LMV226/LMV228 Application with Directional Coupler

SHUTDOWN FUNCTIONALITY

The LMV225/LMV226/LMV228 RF_{IN}/E_N pins have 2 functions combined:

- Enable/Shutdown
- Power input

The capacitor C and the resistor R₂ ([Figure](#page-22-0) 66 and Figure 67) separate the DC shutdown functionality from the AC power measurement. The device is active when Enable = HI, otherwise it is in a low power consumption shutdown mode. During shutdown the output will be LOW.

Capacitor C should be chosen sufficiently large to ensure a corner frequency far below the lowest input frequency to be measured. In case of the LMV225 the corner frequency can be calculated using:

$$
f = \frac{1}{2 \pi (R_1 + R_{IN}) \frac{C \cdot C_{IN}}{C + C_{IN}}}
$$

where

• $R_{\text{IN}} = 50\Omega$, $C_{\text{IN}} = 45$ pF typical (3)

With R_1 = 1800Ω and C = 100 pF, this results in a corner frequency of 2.8 MHz. This corner frequency is an indicative number. The goal is to have a magnitude transfer, which is sufficiently flat in the used frequency range; capacitor C should be chosen significantly larger than capacitor C_{IN} to assure a proper performance of the high resistive tap. Capacitor C shouldn't be chosen excessively large since the RC-time, it introduces in combination with resistor R_2 , adds to the turn-on time of the device.

The LMV226/LMV228 do not use a resistor R₁ like the LMV225. Though a resistor is seen on the coupler side (R_{COUPLER}). Therefore a similar equation holds for the LMV226/LMV228 LF corner frequency, where R_1 is replaced with the coupler output impedance (R_{COUPLER}) .

With R_{COUPLER} = 50 Ω and C = 100 pF, the resulting corner frequency is 50 MHz.

The output voltage is proportional to the logarithm of the input power, often called "linear-in-dB". [Figure](#page-23-0) 68 shows the typical output voltage versus PA output power of the LMV225 setup as depicted in [Figure](#page-21-0) 66.

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Figure 68. Typical power detector response, V_{OUT} vs. PA output Power

OUTPUT RIPPLE DUE TO AM MODULATION

A CDMA modulated carrier wave generally contains some amplitude modulation that might disturb the RF power measurement used for controlling the PA. This section explains the relation between amplitude modulation in the RF signal and the ripple on the output of the LMV225/LMV228. Expressions are provided to estimate this ripple on the output. The ripple can be further reduced by lowpass filtering at the output. This is realized by connecting an capacitor from the output of the LMV225/LMV228 to ground.

Estimating Output Ripple

The CDMA modulated RF input signal of [Figure](#page-23-0) 68 can be described as:

 $V_{IN}(t) = V_{IN} [1 + \mu(t)] \cos (2 \cdot π \cdot f \cdot t)$

where

- V_{IN} is the amplitude of the carrier frequency
- Amplitude modulation $\mu(t)$ can be between -1 and 1 (4)

Figure 69. AM Modulated RF Signal

The ripple observed at the output of the detector equals the detectors response to the power variation at the input due to AM modulation ([Figure](#page-23-1) 69). This signal has a maximum amplitude $V_{\text{IN}} \cdot (1+\mu)$ and a minimum amplitude $V_{\text{IN}} \cdot (1-\mu)$, where 1+μ can be maximum 2 and 1-μ can be minimum 0. The amplitude of the ripple can be described with the formula:

where

- V_Y is the slope of the detection curve ([Figure](#page-24-0) 70)
- μ is the modulation index (5)

[Equation](#page-24-1) 5 can be reduced to:

$$
V_{\text{RIPPLE}} = V_Y \cdot 20 \text{ LOG} \left[\frac{1 + \mu}{1 - \mu} \right]
$$

(6)

Consequently, the ripple is independent of the average input power of the RF input signal and only depends on the logarithmic slope V_Y and the ratio of the maximum and the minimum input signal amplitude.

For CDMA, the ratio of the maximum and the minimum input signal amplitude modulation is typically in the order of 5 to 6 dB, which is equivalent to a modulation index μ of 0.28 to 0.33.

A further understanding of the equation above can be achieved via the knowledge that the output voltage V_{OUT} of the LMV225/LMV228 is linear in dB, or proportional to the input power P_{IN} in dBm. As discussed earlier, CDMA has a modulation in the order of 5 to 6 dB. Since the transfer is linear in dB, the output voltage V_{OUT} will vary linearly over about 5 to 6 dB in the curve [\(Figure](#page-24-0) 70).

Figure 70. V_{OUT} vs. RF Input Power P_{IN}

The output voltage variation ΔV_{OUT} is thus identical for RF input signals that fall within the linear range (in dB) of the detector. In other words, the output variation is independent of the absolute RF input signal:

$$
\Delta V_{\rm O} = V_{\rm Y} \cdot \Delta P_{\rm IN} \tag{7}
$$

In which V_y is the slope of the curve. The log-conformance error is usually much smaller than the ripple due to AM modulation. In case of the LMV225/LMV228, $V_Y = 40$ mV/dB. With $\Delta P_{IN} = 5$ dB for CDMA, $\Delta V_{OUT} = 200$ mV_{PP} . This is valid for all V_{OUT} .

Output Ripple with Additional Filtering

The calculated result above is for an unfiltered configuration. When a low pass filter is used by shunting a capacitor of e.g. C_{OUT} = 1.5 nF at the output of the LMV225/LMV228 to ground, this ripple is further attenuated. The cut-off frequency follows from:

$$
f_C = \frac{1}{2 \pi C_{OUT} R_O}
$$
 (8)

With the output resistance of the LMV225/LMV228 R_O = 19.8 kΩ typical and C_{OUT} = 1.5 nF, the cut-off frequency equals $f_C = 5.36$ kHz. A 100 kHz AM signal then gets attenuated by 5.36/100 or 25.4 dB. The remaining ripple will be less than 20 mV. With a slope of 40 mV/dB this translates into an error of less than \pm 0.5 dB. Since the LMV226 has a low output impedance buffer, a capacitor to reduce the ripple will not be effective.

Output Ripple Measurement

[Figure](#page-25-0) 71 shows the ripple reduction that can be achieved by adding additional capacitance at the output of the LMV225/LMV228. The RF signal of 900 MHz is AM modulated with a 100 kHz sinewave and a modulation index of 0.3. The RF input power is swept while the modulation index remains unchanged. Without the output capacitor the ripple is about 200 mV_{PP}. Connecting a capacitor of 1.5 nF at the output to ground, results in a ripple of 12 mV_{PP}. The attenuation with a 1.5 nF capacitor is then 20 • log (200/12) = 24.4 dB. This is very close to the calculated number of the previous paragraph.

Figure 71. Output Ripple vs. RF Input Power

PRINCIPLE OF OPERATION

The logarithmic response of the LMV225/LMV226/LMV228 is implemented by a logarithmic amplifier as shown in [Figure](#page-25-1) 72. The logarithmic amplifier consists of a number of cascaded linear gain cells. With these gain cells, a piecewise approximation of the logarithmic function is constructed.

Figure 72. Logarithmic Amplifier

Every gain cell has a response according to [Figure](#page-26-0) 73. At a certain threshold (E_K) , the gain cell starts to saturate, which means that the gain drops to zero. The output of gain cell 1 is connected to the input of gain cell 2 and so on.

Figure 73. Gain Cell

All gain cell outputs are AM-demodulated with a peak detector and summed together. This results in a logarithmic function. The logarithmic range is about:

 $20 \cdot n \cdot log(A)$

where

- $n =$ number of gain cells
- $A =$ gain per gaincell (9)

[Figure](#page-26-1) 74 shows a logarithmic function on a linear scale and the piecewise approximation of the logarithmic function.

Figure 74. Log-Function on Lin Scale

[Figure](#page-26-2) 75 shows a logarithmic function on a logarithmic scale and the piecewise approximation of the logarithmic function.

Figure 75. Log-Function on Log Scale

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The maximum error for this approximation occurs at the geometric mean of a gain section, which is e.g. for the third segment:

$$
\sqrt{\frac{E_K}{A^2} \cdot \frac{E_K}{A^1}} = \frac{E_K}{A \sqrt{A}}
$$

(10)

The size of the error increases with distance between the thresholds.

LAYOUT CONSIDERATIONS

For a proper functioning part a good board layout is necessary. Special care should be taken for the series resistance R₁ ([Figure](#page-21-0) 66) that determines the attenuation. For high resistor values the parasitic capacitance of the resistor may significantly impact the realized attenuation. The effective attenuation will be lower than intended. To reduce the parasitic capacitance across resistor R_1 , this resistor can be composed of several components in series instead of using a single component.

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REVISION HISTORY

PACKAGING INFORMATION

(1) The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

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RoHS Exempt: TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

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(3) MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

⁽⁵⁾ Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

(6) Lead/Ball Finish - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead/Ball Finish values may wrap to two lines if the finish value exceeds the maximum column width.

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PACKAGE MATERIALS INFORMATION

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TAPE AND REEL INFORMATION

QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE

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PACKAGE MATERIALS INFORMATION

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PACKAGE OUTLINE

YPD0004 DSBGA - 0.395 mm max height

DIE SIZE BALL GRID ARRAY

NOTES:

- 1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
- 2. This drawing is subject to change without notice.

EXAMPLE BOARD LAYOUT

YPD0004 DSBGA - 0.395 mm max height

DIE SIZE BALL GRID ARRAY

NOTES: (continued)

3. Final dimensions may vary due to manufacturing tolerance considerations and also routing constraints. See Texas Instruments Literature No. SNVA009 (www.ti.com/lit/snva009).

EXAMPLE STENCIL DESIGN

YPD0004 DSBGA - 0.395 mm max height

DIE SIZE BALL GRID ARRAY

NOTES: (continued)

4. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release.

B. This drawing is subject to change without notice.

MECHANICAL DATA

NGF0006A

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